

# ME-221

## SOLUTIONS FOR PROBLEM SET 8-9

### Problem 1

The impulse response of the system is:

$$g(t) = 2\varepsilon(t)e^{-2t}$$

a) The corresponding transfer function of the system is:

$$G(s) = \mathcal{L}[g(t)] = \frac{2}{s+2} \quad (1)$$

b) The unit step response ( $u(t) = \varepsilon(t)$ ) is:

$$Y(s) = G(s)U(s) = \frac{2}{(s+2)} \frac{1}{s} = \frac{1}{s} - \frac{1}{s+2}$$

Therefore,

$$\gamma(t) = y(t) = \mathcal{L}^{-1}[Y(s)] = \varepsilon(t) - \varepsilon(t)e^{-2t} = \varepsilon(t)[1 - e^{-2t}]$$

c) We have a first order system with no zeros. Transfer function in equation 1 can be written in the following form:

$$G(s) = \frac{Y(s)}{U(s)} = \frac{K}{\tau s + 1} \quad (2)$$

where the time constant  $\tau = 0.5$  and the steady-state gain  $K = 1$ .

d) For all systems with a transfer function which can be written as equation 2, the time it takes for the output to reach 95% of its steady-state value is:

$$t = 3\tau = 1.5 \text{ s}$$

e) The Laplace transform of the input  $u(t) = \varepsilon(t-1)e^{-(t-1)}$  is obtained using the temporal translation property:

$$U(s) = e^{-s} \mathcal{L}[\varepsilon(t)e^{-t}] = \frac{e^{-s}}{s+1}$$

The response is:

$$y(t) = \mathcal{L}^{-1}[G(s)U(s)] = \mathcal{L}^{-1}\left[\frac{2}{(s+2)} \frac{e^{-s}}{(s+1)}\right] = \mathcal{L}^{-1}[Y_1(s)e^{-s}]$$

We will first consider the term  $Y_1(s) = \frac{2}{(s+2)(s+1)}$  as the term  $e^{-s}$  will only delay the response  $y_1(t)$  of a unit time.

$$Y_1(s) = \frac{2}{(s+2)(s+1)} = \frac{2}{s+1} - \frac{2}{s+2}$$

Therefore:

$$y_1(t) = 2e^{-t} - 2e^{-2t} \quad t \geq 0$$

and:

$$y(t) = y_1(t-1) = 2e^{-(t-1)} - 2e^{-2(t-1)} \quad t \geq 1$$

## Problem 2

a) The Laplace transform of the system is:

$$Y(s)[s^2 + ks + 4] = U(s) \quad (3)$$

$$G(s) = \frac{Y(s)}{U(s)} = \frac{1}{s^2 + ks + 4} \quad (4)$$

It is a second-order system without zeros, which can be written in the form:

$$G(s) = K \frac{\omega_0^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2}$$

Rewriting equation 4 we get:

$$G(s) = \frac{1}{4} \frac{4}{s^2 + ks + 4} \quad (5)$$

with the steady state gain  $K = 1/4$ , natural frequency  $\omega_0 = \sqrt{4} = 2$ ,  $2\zeta\omega_0 = k$  and therefore damping coefficient  $\zeta = k/4$ .

**b)** The poles of the system are:

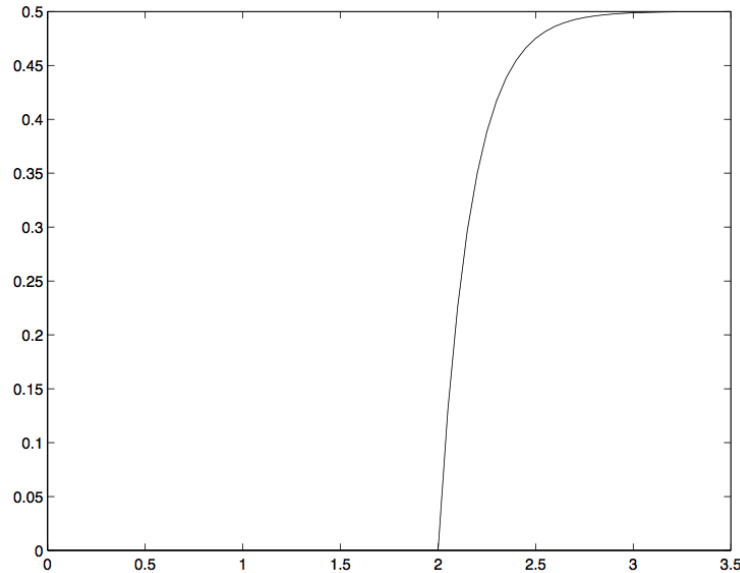
$$p_{1,2} = -\omega_0(\zeta \pm \sqrt{\zeta^2 - 1}) = -\frac{k}{2} \pm \sqrt{\frac{k^2}{4} - 4}$$

The type of the response will depend on the value of  $k$ :

- For  $k < 0$ , we have that  $\text{Re}\{p_{1,2}\} > 0$ . Because of this, the system is unstable as the response includes *increasing* exponential terms. Also,  $\text{Im}\{p_{1,2}\} \neq 0$ , therefore the system will be oscillatory.
- For  $k = 0$ , we have that  $\text{Re}\{p_{1,2}\} = 0$ , and  $\text{Im}\{p_{1,2}\} \neq 0$ , the system will be undamped. In other words, it will never converge nor diverge (because of the real part) but will keep oscillating (because of the imaginary part).
- For  $0 < k < 4$ , we have that  $\text{Re}\{p_{1,2}\} < 0$ , leading to a stable system.  $\zeta < 1$ , therefore, the system will be underdamped. There will be an oscillatory motion since  $\text{Im}\{p_{1,2}\} \neq 0$ .
- For  $k = 4$ , we have that  $\text{Re}\{p_{1,2}\} < 0$ , and  $\zeta = 1$  (there will be one repeated pole:  $p_1 = p_2$ ). Therefore, the system will be stable and critically damped. There will be no oscillatory motion since  $\text{Im}\{p_{1,2}\} = 0$ .
- For  $k > 4$ , we have that  $\text{Re}\{p_{1,2}\} < 0$ , and  $\zeta > 1$  Therefore, the system will be stable and overdamped. There will be no oscillatory motion since  $\text{Im}\{p_{1,2}\} = 0$ .

## Problem 3

The unit-step response is here plotted:



a) The graph shows the output response of a first-order system delayed by 2 seconds. The transfer function will hence be:

$$G(s) = \frac{K e^{-2s}}{\tau s + 1}$$

The steady state gain  $K$  is the ratio between the final output value (0.5) and the amplitude of the input step function (1). Therefore,  $K = 0.5$ .

The time constant  $\tau$ , corresponds to the time frame between the system's first reaction ( $t = 2s$ ) and the moment in which the response reaches 63% of its final value ( $t = 2.167s$ ). Hence,  $\tau = 0.167s$ .

$$G(s) = \frac{0.5e^{-2s}}{0.167s + 1} = \frac{3e^{-2s}}{s + 6}$$

b)

$$g(t) = \mathcal{L}^{-1}[G(s)] = 3\varepsilon(t - 2)e^{-6(t-2)}$$

## Problem 4

a) We know that the transfer function of the system is:

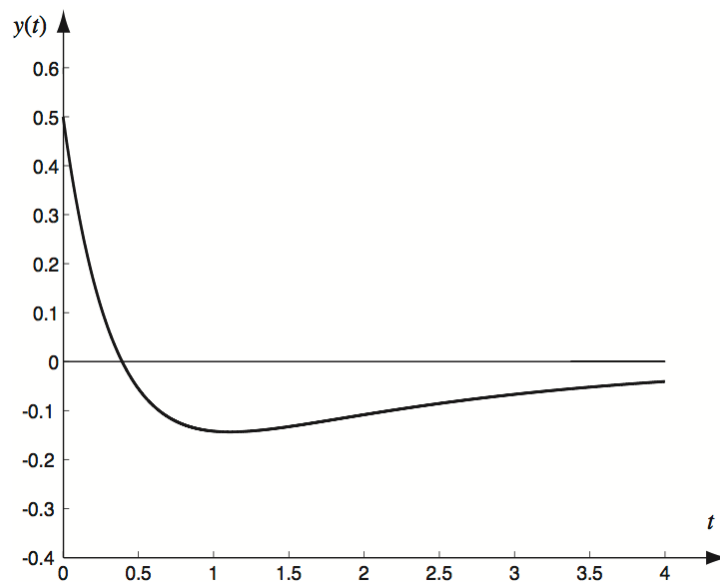
$$G(s) = \frac{s - 1}{(2s + 1)(s + 3)}$$

The impulse response of the system is therefore:

$$\begin{aligned} Y(s) &= G(s)U(s) = \frac{s - 1}{(2s + 1)(s + 3)} = \frac{A}{2s + 1} + \frac{B}{s + 3} \\ \rightarrow A &= \lim_{s \rightarrow -1/2} \frac{s - 1}{s + 3} = -3/5 \\ \rightarrow B &= \lim_{s \rightarrow -3} \frac{s - 1}{2s + 1} = 4/5 \end{aligned}$$

The impulse response is hence:

$$y(t) = \mathcal{L}^{-1}Y(s) = -\frac{3}{10}e^{-t/2} + \frac{4}{5}e^{-3t} \quad t \geq 0$$



b)  $z_1 = 1$ ,  $p_1 = -0.5$  and  $p_2 = -3$ . Both poles are in the left quadrants, the system is stable.

## Problem 5

a) The value of the unit step response at peak time is given by:

$$\gamma(t_p) = K(1 + e^{-\zeta\pi/\sqrt{1-\zeta^2}})$$

From the plot, we can read the static gain, which is  $K = 2$ . The peak value is recorded as  $\gamma(t_p) = 3$  in the plot. We can extract the value of the damping coefficient  $\zeta$  as follows:

$$\begin{aligned}\gamma(t_p) &= 2(1 + e^{-\zeta\pi/\sqrt{1-\zeta^2}}) = 3 \\ \rightarrow \zeta &= \frac{\ln 2}{\sqrt{\pi^2 + (\ln 2)^2}} = 0.215\end{aligned}$$

We can calculate  $\omega_0$  from the value of  $t_0$ :

$$t_p = \frac{\pi}{\bar{\omega}} = \frac{\pi}{\omega_0 \sqrt{1-\zeta^2}} = 1 \rightarrow \omega_0 = \frac{\pi}{\sqrt{1-\zeta^2}} = 3.217$$

b) The transfer function is therefore:

$$G(s) = Ke^{-s} \frac{\omega^2}{s^2 + 2\zeta\omega_0 s + \omega_0^2} = \frac{20.6e^{-s}}{s^2 + 1.38s + 10.3}$$

(Note 1s delay)

## Problem 6

a) Input: force  $F(t)$ , Output: position  $x(t)$ . We can then write the mathematical model of the Newton's law for  $t \geq 0$ :

$$(m + M)\ddot{x} = Mg - kx - f\dot{x}$$

With:  $x(0) = \dot{x}(0) = 0$

Remark:

Left side of the equation  $\rightarrow$  both masses ( $m + M$ ) are accelerated.

Right side of the equation  $\rightarrow$  only  $M$  is taken into account as  $m$  is already taken into account in the definition  $x = 0$  for  $t \leq 0$ .

b) This system is rather particular as the application of the input modifies one of parameter of the system ( $m$  for  $t < 0$  and  $m + M$  for  $t \geq 0$ ). However, for  $t \geq 0$ , the mass of the system remains constant.

The Laplace transform of the system is:

$$\begin{aligned}[(m + M)s^2 + fs + k]X(s) &= F(s) \\ G(s) = \frac{X(s)}{F(s)} &= \frac{1}{(m + M)s^2 + fs + k} = \frac{1}{100s^2 + 350s + 3500}\end{aligned}$$

It is a second-order system without zeros. Its poles are:

$$100s^2 + 350s + 3500 = 2s^2 + 7s + 70 = 0$$

$$p_{1,2} = \frac{-7 \pm \sqrt{49 - 560}}{4} = -1.75 \pm 5.65j$$

c) As it is a second-order system, the denominator can be written as:  $\tau^2 s^2 + 2\zeta\tau s + 1$

Hence,

$$\frac{1}{35}s^2 + \frac{1}{10}s + 1$$

With:  $\tau = 0.169s$ , and  $\zeta = 0.269$ . This corresponds to an highly oscillatory response.

d) The response is obtained by:

$$X(s) = G(s)F(s) = \frac{1}{100s^2 + 350s + 3500}(90)(9.81)\frac{1}{s}$$

$$= \frac{17.66}{s(2s^2 + 7s + 70)} = \frac{A}{s} + \frac{Bs + C}{2s^2 + 7s + 70}$$

Where:  $A = 0.252$ ,  $B = -0.504$  and  $C = -1.766$ .

By looking at the laplace transform tables, the denominator informs us on which form we should have it to enable us to easily find the inverse Laplace. In this case:

$$2s^2 + 7s + 70 = 2(s^2 + 3.5s + 35)$$

resembles the following denominator from the Laplace tables:

$$(s + \alpha)^2 + \omega^2 = s^2 + 2\alpha s + \alpha^2 + \omega^2$$

We therefore get:

$$2\alpha = 3.5$$

$$\alpha^2 + \omega^2 = 35$$

We find that:  $\alpha = 1.75$  and  $\omega = 5.65$ .

The next step is to work on the numerator. With what we found above we have:

$$\frac{Bs + C}{2s^2 + 7s + 70} = \frac{\frac{B}{2}s + \frac{C}{2}}{s^2 + 3.5s + 35} = \frac{\frac{B}{2}s + \frac{C}{2}}{(s + \alpha)^2 + \omega^2} = \frac{-0.252s - 0.883}{(s + \alpha)^2 + \omega^2}$$

The numerator should also be put into a form similar to that found in the Laplace table. In this case, we can see that we can write it as a linear combination of  $(s + \alpha)$  and  $\omega$ :

$$-0.252s - 0.833 = c_1(s + \alpha) + c_2\omega$$

We get:

$$\begin{aligned}c_1 &= -0.252 \\ -0.883 &= -0.252\alpha + c_2\omega \Rightarrow c_2 = -0.078\end{aligned}$$

Therefore, we have:

$$\frac{Bs + C}{2s^2 + 7s + 70} = -0.252 \frac{(s + \alpha)}{(s + \alpha)^2 + \omega^2} - 0.078 \frac{\omega}{(s + \alpha)^2 + \omega^2}$$

The inverse Laplace transform of  $X(s)$  gives:

$$x(t) = 0.252[1 - e^{-1.75t}\cos 5.65t - 0.31e^{-1.75t}\sin 5.65t] \quad t \geq 0$$